



A time domain transmission sensor with TDR performance characteristics

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Abstract

Time domain reflectometry (TDR) has become a standard method for determining water content in soils. However, the cost of instrumentation and the level of user-ability required often limit its application, especially for agricultural and urban soil water management. The objective of this paper is to report on a new, lower cost time domain transmission (TDT) sensor and to demonstrate that it has the performance characteristics of commonly used TDRs. We approximate the sampling volume of the TDT sensor and compare it to that of TDR. We demonstrate that the sensor accurately determines permittivity from travel time measurements by comparing the TDT sensor with two TDR instruments, and a network analyzer providing measurements of frequency-dependent permittivity. Both TDRs and the TDT operated within ±3 permittivity units of each other across the permittivity range of 9–80. The reported rise time of the TDT sensor is 180 ps, which suggests a frequency bandwidth and upper frequency limit similar to TDR. We determined the maximum passable frequencies of all three instruments in non-relaxing media and demonstrate that they are in fact quite similar with the average maximum passable frequencies of the two TDR systems being 1.64 and 1.45 GHz, and the average maximum passable frequency of the TDT being 1.23 GHz. In addition to having accuracy and frequency characteristics similar to TDR, the TDT sensor offers the advantage of having the pulse generating and sampling electronics mounted in the head of the probe which reduces attenuation and overcomes the constraint that prevents TDR being used with long coaxial cables and multiplexers connecting the instrument to TDR probes.

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1. Introduction

Transmission line, electromagnetic measurement methods have proved reliable for estimating soil water content in both laboratory and field settings (Topp and Ferre, 2002; Robinson et al., 2003a). Understanding of both the measurement technique (Topp and Ferre, 2002; Robinson et al., 2003a) and the dielectric properties of soil (Heimovaara et al., 1994; Friedman, 1998) have advanced considerably since the seminal paper presented by Topp et al. (1980). In many water management applications, water content sensor technology such as TDR could greatly improve efficiency and conserve energy and water resources. TDR is not the solution to all soil

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water content sensing problems, but has the advantage of being in-situ, real-time, and more accurate than less expensive, low frequency devices. TDR sensing utilizes probes that are relatively easy to manufacture and insert into samples. Ease of insertion is advantageous in situations where minimum sample disturbance is required. TDR can also simultaneously determine water content and electrical conductivity within certain limits. In addition, automated collection and the instrument's ability to be connected with a multiplexer allows for a number of locations to be sampled, almost simultaneously.

1.1. TDR

Despite all of it's advantages, the cost of TDR and the level of ability required by the operator often places it beyond the means of growers. Within the last few years, there have been a number of attempts to address many of these issues. Systems such as the TDR100 (Campbell Scientific, Logan, UT; www. campbellsci.com/tdr100 verified 14 April 2005), Mini Trase (Soil Moisture Equipment Corps, Santa Barbara, CA; www.soilmoisture.com verified 14 April 2005), and Trime (Mesa Systems Co., Medfield, MA; www.mesasystemsco.com verified 14 April 2005) are smaller, more portable and more rugged (Robinson et al., 2003a,b). However, the design concept of a stand-alone sensor of low cost (<\$500), small size, high accuracy, and precision in the determination of permittivity that covers a representative sampling volume has not been forthcoming with TDR technology. The closest instrument is perhaps the Theta Probe (Dynamax Inc., Houston, TX; http//:www. dynamax.com verified 14 April 2005), which uses transmission line technology to measure impedance (Gaskin and Miller, 1996). Capacitance and impedance probes have tended to fill the lower price market. These instruments tend to be limited to operating frequencies < 150 MHz which is undesirable if the soil has dielectric dispersion in this frequency range, which is mostly attributed to 2:1 clay minerals such as illite or montmorillonite. Most of these instruments do not permit the measurement of bulk soil electrical conductivity, which can be useful for management purposes. Many of the capacitance sensors are sensitive to interference from bulk soil electrical conductivity (Robinson et al., 1998; Kelleners et al., 2004) and while many will continue to operate, the prediction of water content can be very poor (Baumhardt et al., 2000). In addition to cost and user-ability limitations, the length of cable that the probe can be attached to without signal attenuation compromising accurate permittivity and subsequently, water content estimations, limits TDR sensor systems. Sensor systems in field site locations must have coaxial cable running from the probe to the cable tester (i.e. signal generating and sampling electronics), which with signal attenuation in the cables tends to limit the radial distance that probes can be placed to about 30-m. Signal attenuation in cables filters out the higher frequency signal components in the same manner that dielectric relaxation of the sampled medium filters the higher frequency signal components (Robinson et al., 2003a). This is disadvantageous owing to the higher frequency signal components determining the signal travel time.

1.2. Recent developments using TDT

Topp et al. (2001) described the use of TDT as part of a water content sensing penetrometer. They helically wrapped a parallel pair transmission line around the end of a penetrometer and connected the pulse transmitting and sampling hardware via 50 Ω coaxial cables (Topp et al., 2001). They converted the TDT frequency output to voltage via a frequency to voltage converter and related the measured signal voltages to permittivity values via empirical calibration in organic liquids of known permittivity (Topp et al., 2001). The voltage differences between samples result from increasing propagation velocity (i.e. decreasing travel time) along the transmission line as the permittivity of the liquids decreases (Topp et al., 2001). Permittivity values empirically determined (from the permittivity/voltage calibration) in laboratory and field soil samples then allowed for soil water content prediction. Using this method, they reported an error in predicted water content of $\pm 0.02 \,\mathrm{m}^3 \,\mathrm{m}^{-3}$ (Topp et al., 2001). Water content prediction errors were attributed to sample disturbance caused during penetrometer insertion (Topp et al., 2001).

Harlow et al. (2003) applied two TDT techniques to estimate the water content of variably saturated

sands with a wide range of pore water electrical conductivities by connecting a network analyzer to a 2-rod parallel transmission line embedded in a column (i.e. sample container). The first of their methods determined travel time in the sample by calculating the time difference between transmitted pulse peaks (displayed in the time domain) in the sample and in air (Harlow et al., 2003). This method assumes that the pulse travel time along the cable linking the network analyzer and the sample is the same for each measurement. Their second method involved converting time domain measurements to the frequency domain via Fast Fourier Transformation and then calculating travel time in the sample via phase difference between air and the sample (Harlow et al., 2003). They reported that the travel time/ water content calibrations derived using their TDT techniques compare quite well to TDR travel time/ water content calibrations made by Hook and Livingston (1995).

Hook et al. (2004) also employed the same TDT methods used by Harlow et al. (2003) to investigate electrical conductivity effects on the accuracy and uncertainty of TDR water content estimations in saturated sands. They applied TDT measurement methods to negate the large air/soil boundary reflection often associated with TDR measurements (Hook et al., 2004). Their results showed good correlation between pulse rise time, which increased with increasing sample electrical conductivity, and the average error associated with water content predictions (Hook et al., 2004).

Acclima® has developed a TDT soil water content sensor referred to as a Digital TDT Sensor. Similar to TDR, which relies on a reflection travel time measurement, the Digital TDT Sensor estimates dielectric permittivity using a transmission travel time measurement. A major advantage of the Acclima Digital TDT Sensor is a considerably reduced cost and packaging size. The objective of this study was to test the performance of the Digital TDT Sensor against two TDR instruments, comparing probe design, sampling volume, permittivity estimation, and maximum passable frequency. The Acclima Digital TDT Sensor (Acclima Inc., Meridian, ID; http//:www.acclima.com/Products/Sensor verified 14 April 2005) was compared with the Tektronix 1502B cable tester (Tektronix Inc., Beaverton, OR; http//:www.tektronix.com verified 14 April 2005) and the Campbell Scientific TDR100 both connected to a standard 0.15-m probe.

2. Theoretical considerations

2.1. Transmission line theory

TDR and TDT transmission line techniques can be used to estimate dielectric permittivity from the travel time of an electromagnetic signal propagating along a probe buried in a dielectric, which could be any liquid or porous medium. The phase velocity (v_p) of the instrument signal is a function of the electromagnetic properties of the medium

$$v_{\rm p} = \frac{c}{\sqrt{\varepsilon_{\rm r}\mu_{\rm r}}}\tag{1}$$

where c is the speed of light in vacuum (3×10⁸ m s⁻¹), and ε_r and μ_r are the dielectric permittivity and magnetic permeability of the medium, relative to vacuum. Most materials are non-magnetic, thus μ_r is equal to one and ε_r determines the signal propagation velocity. TDR estimates an apparent permittivity (K_a) which is considered equivalent to ε_r for materials without dielectric loss. K_a is determined by rearranging Eq. (1) for ε_r (= K_a) and setting v_p =L/t as follows:

$$K_{\rm a} = \left(\frac{c}{v}\right)^2 = \left(\frac{ct}{2L}\right)^2 \tag{2}$$

where L is the physical probe length (m) and t is the pulse travel time (s) in the sample. With TDR the pulse is reflected at the end of the probe and the return signal is sampled. The factor 2 in the denominator of Eq. (2) accounts for the two-way (down and back) travel time of the TDR signal. With TDT the pulse travels the length of the probe once and the transmitted signal is sampled. Thus, TDT determination of K_a is the same as with TDR, the only difference being measurement of one-way travel time (i.e. the factor of 2 is omitted from Eq. (2)). For most hydrological and environmental applications the determination of K_a leads to a calculation of volumetric water content (θ_{v}) made using either empirical equations (Topp et al., 1980; Malicki et al., 1996) or dielectric mixing models (Roth et al., 1990; Dirksen and Dasberg, 1993; Friedman, 1998).

2.2. Permittivity measurement and water content determination

It has been established that for many soils there is a strong relationship between K_a , estimated by transmission line sensors, and θ_v (Topp et al., 1980; Malicki et al., 1996). The reason for this is the strong contrast between the permittivities (ε) of water ($\varepsilon_{\rm w} \approx 80$) with those of mineral soil solids ($\varepsilon_s \approx 2-9$) and air ($\varepsilon_a \approx 1$). The accuracy and precision of dielectric permittivity determined using TDR enables accurate θ_v predictions (less than $\pm 0.02 \,\mathrm{m}^3 \,\mathrm{m}^{-3}$) in many coarse textured soils (Topp et al., 1980; Hook and Livingston, 1995). Although estimating $\theta_{\rm v}$ is generally the ultimate goal, it is inferred from K_a estimates using travel time analysis, which is subject to other secondary factors including water status (e.g. bound, free), particle shape, configuration of constituents (e.g. water, solid, air), ion concentration, and others (Jones and Or, 2002, 2003). The effect on the estimated permittivity may be related to unwanted relaxations (e.g. Maxwell-Wagner) or other porous media-related effects that alter the apparent water-phase permittivity leading to potential errors in water content determination. For this reason, the instrument comparisons herein utilize liquids that provide homogeneous backgrounds and uniform dielectrics rather than soils or other porous media which tend to enhance unwanted noise and uncertainty in the measurements.

3. Materials and methods

3.1. TDR and TDT probe description and design

Transmission line sensor K_a estimate quality is largely dependent on probe design (Heimovaara, 1993; Ferre et al., 1998; Robinson et al., 2003a). Probes are designed to imitate a coaxial transmission line while being versatile enough for both laboratory and field applications. Optimum probe design attempts to maximize the representative sampling volume, or the sampled porous medium volume which contributes to the measurement (Ferre et al., 1998), while at the same time minimizing signal attenuation across the rods and allowing for easy insertion into samples. Fig. 1 shows a photograph of the Campbell Scientific TDR100 connected to

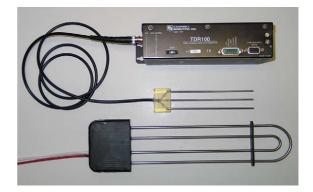


Fig. 1. Photograph comparing the Acclima Digital TDT Sensor and the Campbell Scientific TDR100 connected to a conventional 0.15-m 3-rod probe.

a conventional 0.15-m 3-rod probe and the Acclima Digital TDT Sensor (0.60-m 2-rod loop probe with the electronics contained in the probe head). The cross-section of the Digital TDT Sensor provides a much greater sampling area due to its larger probe size and spacing as well as the dual path for the signal travel required using TDT (Fig. 1; Table 1 lists the dimensions of probes).

The electronic components of the Digital TDT Sensor are all contained within the $9.0 \times 7.5 \times 2.5$ cm³

Table 1 Sensor characteristics

	Acclima TDT	Tektronix TDR	CSI TDR100
Length (m)	0.60	0.15	0.15
Electrode	14.2	12.0	12.0
spacing (mm)			
Electrode	4.80	3.20	3.20
diameter			
(mm)			
Output	0.25	0.30	0.25
voltage (V)			
Rise time (ps)	180	200	270
Digitized	1024	251	20-2048 ^a
waveform			
points			
TDT/TDR	\$349	\$11,695	\$3650
cost			
Output device	RS500	Analysis	CR10X data
	Controller	package	logger
Minimum	\$349	\$11,695	\$4840
system cost			
(mm) Output voltage (V) Rise time (ps) Digitized waveform points TDT/TDR cost Output device Minimum	180 1024 \$349 RS500 Controller	200 251 \$11,695 Analysis package	270 20–2048 ^a \$3650 CR10X data logger

^a The number of digitized waveform points measured by the TDR100 is adjustable.

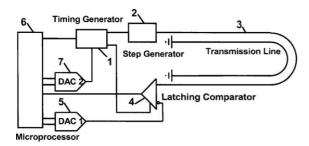


Fig. 2. Schematic drawing of the Acclima Digital TDT Sensor displaying the basic components (Anderson, 2003). The numbers correspond to the following: (1) timing generator; (2) step generator; (3) stainless steal transmission lines; (4) latching comparator; (5) digital to analog converter (DAC) 1; (6) microprocessor; (7) DAC 2.

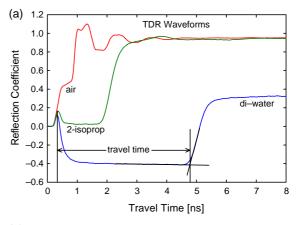
probe head (Fig. 2; Anderson, 2003). The basis for a measurement is briefly described here. Similar to TDR, the transmitting end of the Digital TDT Sensor propagates a fast-rising step function down the transmission line comprised of stainless steel rods. The digital acquisition system is positioned at the receiving end of the transmission line, which digitizes the waveform. The resulting digitized waveform is analyzed by taking the first and second derivatives and extracting maximum slope points and maximum inflection points in the waveform. Proprietary digital signal processing algorithms extract the true propagation time of the received signal and permittivity is calculated from travel time (Eq. (2)). For details on TDR system operation see Robinson et al. (2003a). The rise times, output voltages, and digitized waveform points of the Digital TDT Sensor, Tektronix 1502B TDR, and Campbell Scientific TDR100 are similar (Table 1).

3.2. Waveform interpretation

Both TDR and TDT instruments rely on computer software or firmware that capture and analyze waveforms characteristic of the medium in which the probes are embedded. As explained above, the Acclima Digital TDT Sensor utilizes custom firmware within the probe head to interpret waveforms for travel time. Currently a custom controller is required to power the sensor and retrieve the measured waveforms and data. However, with a modification in the communication protocol via the firmware,

sensor control and data collection with common data loggers is possible. The Tektronix 1502B TDR waveforms were captured and interpreted for travel time measurement with WinTDR waveform analysis software (Or et al., 2003; available at http://129.123. 13.101/soilphysics/wintdr/index.htm verified 14 April 2005). The Campbell Scientific TDR100 measurements were made with Campbell Scientific PCTDR software (included with purchase of TDR100).

TDR and TDT waveforms measured in air, 2-isopropoxyethanol, and water are shown (Fig. 3) for comparison. It should be noted that currently the Acclima Digital TDT Sensor only allows for a maximum voltage value of 0.6225 mV to be output, thus the waveforms shown are terminated everywhere this voltage value is exceeded. Tangent line fitting to



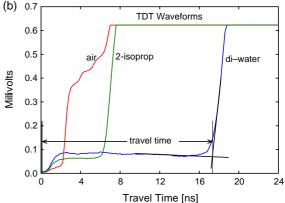


Fig. 3. (a) TDR waveforms in air, de-ionized water, and pure 2-isopropoxyethanol, and (b) TDT waveforms in the same three mediums. The water waveforms display arbitrarily drawn tangent lines to mark the end points from which travel time is measured.

mark the probe ends was accomplished by the described firmware for the Digital TDT Sensor, and WinTDR and PCTDR software for the Tektronix TDR and TDR100, respectively. The Digital TDT Sensor firmware and TDR software algorithms measure travel time via tangent line fitting to mark the beginning and end of the respective probes. The waveforms in Fig. 3 display arbitrarily drawn tangent lines marking the probe ends to illustrate what the firmware and software accomplish. Calibration of both the TDR probe and the Digital TDT Sensor was conducted using measurements in air and de-ionized water according to the methodology described in Heimovaara (1993) and Robinson et al. (2003b). This calibration procedure determines the position of the vertical line marking the beginning of the probe (Fig. 3). Travel times measured with the described firmware and software were adjusted based on this calibration procedure.

3.3. Comparison of sampling volume and permittivity measurement

Knight (1992) related the energy density distribution surrounding TDR probe rods to the sampling volume in a homogeneous isotropic dielectric medium. The relative electromagnetic energy density distribution of both the TDT and TDR probes was modeled using the Arbitrary Transmission Line Calculator, ATLC (http://atlc.sourceforge.net/ verified 14 April 2005) (Kirkby, 1996). The ATLC software program uses a finite difference approach to calculate characteristic impedances of prescribed geometries assumed to function as transmission lines.

An experiment was also conducted to examine the spatial sensitivity of the Digital TDT Sensor using an air/water interface (i.e. significant dielectric discontinuity). The probe was oriented with the long dimension positioned horizontally and then vertically immersed in de-ionized water with permittivity measurements being made at 0.01-m vertical increments. The initial measurement was made in air with the probe suspended 0.15-m above the de-ionized water and the final measurement was made when the instrument was submersed to a 0.15-m depth in the water.

A comparison of permittivity determination was conducted using air (ε_a =1) and 12 solutions ranging

from $\varepsilon \approx 9-80$ under temperature controlled conditions. Comparisons were made between measurements taken with the Tektronix 1502B TDR and Campbell Scientific TDR100, both connected to the described 0.15-m long 3-rod probe, and with the Acclima Digital TDT Sensor. Frequency domain measurements of permittivity were also made using a HP8752C network analyzer and HP85070B dielectric probe, which served as the reference for dielectric measurements. The sample solutions were made using fractions of de-ionized water mixed into 2-isopropoxyethanol with the permittivity extremes, $\varepsilon \approx 9-80$, coming from pure 2-isopropoxyethanol and undiluted de-ionized water. De-ionized water and 2-isopropoxyethanol were selected to make the solutions because neither exhibits significant dielectric relaxation within the TDR frequency bandwidth, which is reported as 20 kHz to 1.75 GHz (Heimovaara, 1994).

3.4. Maximum passable frequency

The frequency bandwidth associated with the electromagnetic signal generated by TDR and TDT sensors are made up of a broad range of frequencies. The maximum passable frequency is the frequency associated with the fastest traveling portion of the transmitted signal in a dielectric material (Robinson et al., 2003a). Essentially this frequency is the highest signal frequency that is not filtered by the sample or cables, and indicates the upper frequency limit of the sensor for the specific sample measured. The maximum passable frequency for each sensor was determined in each of the described 2-isopropoxyethanol/de-ionized water solutions using the method of Or and Rasmussen (1999), also described in Robinson et al. (2003a). This method matches the $K_{\rm a}$ value determined by the TDR or TDT to the frequency-dependent real permittivity (ε') value measured in the frequency domain with the network analyzer. This method relies on the sample liquid in exhibiting slight dielectric relaxation so that there is a small change in the permittivity as a function of frequency. An average maximum passable frequency for each sensor was also calculated by averaging the maximum passable frequencies determined for each sensor in the permittivity range between $\varepsilon \approx 9-60$.

4. Results

4.1. Instrument sampling volume

The representative sampling volume of a given probe is inferred from calculation of the electromagnetic energy density distribution surrounding the probe rods. The results of the electromagnetic energy density calculations (accomplished with the described ATLC software) for the TDR and TDT probes display a cross-section view of the two probes with the black lines outlining the probe heads to indicate scale and the shaded area representing the calculated electromagnetic energy density surrounding the rods (Fig. 4). The darker the shading the denser the electromagnetic energy, with the darker areas contributing greater weight to the measurement (Fig. 4). The density calculations show the densest electromagnetic energy surrounding the middle rod of the TDR probe and a more even distribution between the two TDT rods when compared with the TDR probe (Fig. 4).

The spatial sensitivity (determined in the submersion experiment) of the TDT sensor in the space adjacent to the *y*- and *z*-dimensions of the probe is shown in Fig. 5. The normalized spatial sensitivity is reported as a numerical factor derived by subtracting from one the measured permittivity divided by the expected permittivity, with the expected permittivity

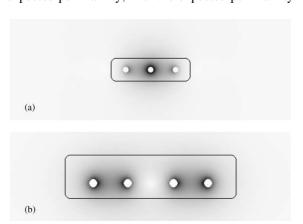


Fig. 4. Cross-sections showing the ATLC (http://atlc.sourceforge.net/verified April 14, 2005) (Kirkby, 1996) modeled electromagnetic energy density surrounding probe rods for (a) conventional 3-rod TDR probe with 3.2-mm diameter rods and 12.0-mm rod spacing, and (b) 2-rod Digital TDT Sensor with 4.8-mm diameter rods and 14.2-mm rod spacing.

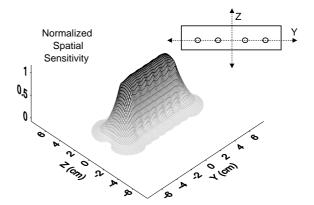
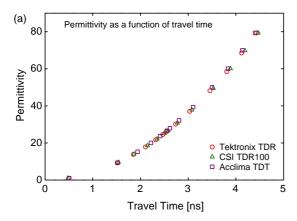


Fig. 5. Three-dimensional plot of the spatial sensitivity of the Digital TDT Sensor in the labeled y- and z-directions showing the distance from a dielectric discontinuity for which the measurement is no longer influenced by the bordering medium (i.e. normalized spatial sensitivity = 0).

being the permittivity measured with the probe completely submersed in a homogenous medium (i.e. water). The spatial sensitivity plot relates to the representative sampling volume and can be interpreted qualitatively as the distance from a significant dielectric discontinuity (i.e. air/water boundary) at which the probe is no longer influenced by the bordering dielectric medium. The Digital TDT Sensor is not greatly influenced by sample that lies beyond approximately 6-cm from the center of the probe in the y-direction and approximately 3-cm from the center of the probe in the z-direction (Fig. 5). The sampling volume will play an important role when considering depth of placement for irrigation management or experimental set-up.

4.2. Permittivity measurements

Comparisons of sensor estimated K_a values as a function of measured travel times are shown in Fig. 6(a). The TDT travel time measurements are divided by a factor of 4 to account for the 0.60-m probe length compared to the 0.15-m TDR probe length. Technically, the Digital TDT Sensor signal travel length is only twice the TDR travel length, but both the Tektronix TDR and TDR100 account for the reflection measurement and output waveforms displaying one-way travel time. Thus, the TDT travel time must be scaled by a factor of 4, despite the fact



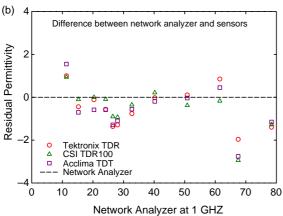


Fig. 6. (a) Permittivity as a function of travel time comparison of the three sensors. In order to account for the scaling issue of the longer TDT probe, the TDT travel times are normalized to the TDR by dividing by a factor of 4. (b) The residual permittivity (sensor estimated K_a value subtracted from the network analyzer ε' value measured at the average maximum passable frequency of the sensor) of each sensor plotted as a function of network analyzer ε' values measured at a frequency of 1 GHz.

that the TDR signal is traveling down and back along the probe. The TDR and TDT sensors follow the same K_a -travel time trend (Fig. 6(a)). The residual permittivities of each sensor were plotted as a function of measured network analyzer ε' values at 1 GHz (Fig. 6(b)). The residual permittivity is the difference between the estimated K_a values for each sensor and network analyzer ε' values measured at the average maximum passable frequency (reported below) of the sensor. All three sensors fall within a range of less than ± 3 permittivity units of the network analyzer (Fig. 6(b)). Measured travel time, and hence K_a ,

differences between the sensors are attributed to the differing average maximum passable frequencies of the sensors and the slight relaxation of the media. The real component of medium permittivity decreases as frequency increases due to dielectric relaxations. Thus, travel time measurements, and hence $K_{\rm a}$ estimates, made at higher frequencies will be slightly lower than those measured at lower frequencies.

4.3. Instrument frequency characteristics

The sensor estimated permittivity values plotted as a function of maximum passable frequency are shown in Fig. 7. The average maximum passable frequencies for the 3-rod 0.15-m probe connected to Tektronix TDR, 3-rod 0.15-m probe connected to the TDR100, and Digital TDT Sensor were 1.64, 1.45, and 1.23 GHz, respectively, and the maximum passable frequency standard deviations were 0.282, 0.222, and 0.369 GHz, respectively. The TDR and TDT maximum passable frequencies compare quite well at all measured data points between the solution permittivity range of $\varepsilon \approx 9-60$ (Fig. 7). This analysis shows that the Digital TDT Sensor frequency bandwidth is comparable to that of TDR. As the instruments all have similar rise times they should all have similar frequency bandwidths. The TDT maximum passable frequency values were generally a little lower than the corresponding TDR measurements. This is considered to be due to the longer signal travel time of the Digital

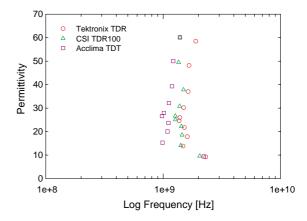


Fig. 7. Maximum passable frequencies of each sensor determined in de-ionized water and 2-isopropoxyethanol mixtures with permittivity values ranging from $\varepsilon \approx 9-60$.

TDT Sensor, which will allow time for slightly more attenuation of the higher frequency components of the signal. As stated above, maximum passable frequency differences explain the slight travel time, and subsequently K_a , differences measured by each of the sensors (Fig. 6(a,b)). Owing to the slight relaxation of the media within the sensor frequency bandwidths, sensors operating at lower a maximum passable frequency will measure slightly higher travel time than sensors operating at a higher maximum passable frequency.

At permittivity values greater than $\varepsilon \approx 60$, it becomes difficult to determine a maximum passable frequency because the solutions show insufficient change in permittivity as a function of frequency to locate a valid value of frequency, thus the two solutions of highest permittivity are omitted from Fig. 7 and not used in the average maximum passable frequency calculations. The maximum passable frequencies for each sensor also vary over the measured K_a range (Fig. 7). The maximum passable frequency differences at each K_a step are attributed to the slight relaxation of the 2-isopropoxyethanol/deionized water solutions within the sensors' frequency bands. The data points corresponding to the lowest maximum passable frequencies are measurements from the solutions showing the most significant relaxation. Dielectric relaxation of the liquid causes more attenuation of the higher frequency components of the signal, and thus lower maximum passable frequency values overall (Robinson et al., 2003a).

5. Discussion

As stated, K_a estimate quality is largely dependent on the sample from which the estimate is taken being representative of the medium. The larger sample volume enclosed by the Digital TDT sensor and the more uniform electromagnetic energy density distribution surrounding the rods (Fig. 4) indicate that the Digital TDT Sensor likely averages travel time measurements from a greater sample volume. More importantly, the uniform electromagnetic energy density distribution of the Digital TDT sensor rods indicates that it also lends more weight to the measurement from a greater sample volume.

In contrast, the density concentration near the middle TDR rod (Fig. 4) gives possibility for increased measurement error because the sample near the middle rod surface is subject to greater disturbance during insertion.

The Tektronix TDR has been considered a standard for $\theta_{\rm v}$ prediction due to its ability to accurately estimate permittivity. The travel time and K_a estimate comparisons (Fig. 6(a,b)) indicate that the Digital TDT Sensor measures travel times and estimates K_a values comparable to the two TDR instruments across the entire permittivity range evaluated here (approximately 9–80), which is commonly measured in soils and porous media. The differences observed between the TDRs and the TDT are likely due to the maximum passable frequency differences of the sensing systems. The Tektronix TDR shows the highest maximum passable frequency, and therefore, the lowest travel times (i.e. fastest traveling signal), while the Digital TDT Sensor shows the lowest maximum passable frequency, and therefore, the highest travel times. As explained, the maximum passable frequency differences reported are likely due to the longer rod length of the Digital TDT Sensor allowing for the signal to experience greater attenuation in the slightly relaxing media used for measurements. It should be remembered that maximum passable frequency is a function of not only the media in which measurements are made and the rod length, but also the cable length connecting the rods to the electronics, which generate and sample the signal. Herein is another advantage of the Digital TDT Sensor over TDR, in that the signal generating and sampling firmware is located within the probe head of the Digital TDT Sensor, thereby negating coaxial cable losses experienced with TDR.

In addition to having the electronics mounted in the probe head and showing measurement accuracy and frequency characteristics similar to TDR, the Acclima Digital TDT Sensor technology can significantly reduce the price constraints of TDR. The costs of the sensor systems considered herein are listed (Table 1) with the minimum system cost representing the minimum cost requirement to make K_a estimates and θ_v predictions. The Digital TDT Sensor was designed to control irrigation according to a threshold water content estimated in the given medium, and thus currently requires a custom controller to power

the sensor and retrieve the measured data. Different controllers currently exist with the cheapest of these being the RS500, whose price is listed (Table 1). As explained above, with a modification in the communication protocol via the firmware the possibility of a stand-alone probe costing \$349 (Table 1) and outputting water content directly to a data logger exists.

The major disadvantage with TDT technology is that it requires a sensing loop (Fig. 1), which generally means soil excavation rather than insertion for sensor placement. This could potentially be a source of error owing to the possibility of introducing density differences between the sample surrounding the rods and the rest of the medium being characterized. In addition, the longer rod length (0.60-m) of the Digital TDT Sensor allows for greater attenuation of the signal, and therefore, more filtering of the higher frequency components of the signal. The longer probe length was selected to maintain timing accuracy and optimize the trade-off between resolutions and signal attenuation, both of which increase with increasing length. However, assuming the rods are the same length for given TDR and TDT probes, the TDT would experience less signal attenuation owing to it's one-way travel time. Also, the possibility exists of manufacturing probes with shorter rods and converting the current TDT measurement to a TDR measurement, which would allow for pointed rods.

While the Digital TDT Sensor is currently designed to control irrigation, the high quality measurement capability of the sensor coupled with the unique design of installing the firmware within the probe head and low cost heralds a new generation of $\theta_{\rm v}$ sensing technology that will advance water management and sensing capabilities. Sensors of this nature are particularly suited not only to developing water savings in turf grass management and irrigation scheduling for gardens and municipal areas, but would provide greater accuracy in weather station monitoring of soil moisture which is a growing need for remote sensing measurement validation. We see the Acclima Digital TDT Sensor potentially providing a rugged alternative to TDR in precision laboratory instrumentation applications in addition to a variety of hydrologic and water management applications.

6. Conclusions

TDR is widely accepted as a standard real-time, in-situ technique for estimating soil water content owing to its ability to make relatively accurate permittivity estimates and to the exploitation of the significant permittivity contrast between water and other porous medium constituents. However, TDR applications may be limited due to high costs, userability requirements, and problems when connecting TDR probes to long lengths of cable. The Acclima Digital TDT sensor has the potential to offer a more affordable alternative. The Acclima Digital TDT Sensor frequency bandwidth and permittivity estimates based on travel time measurements compare quite well to those of the Tektronix TDR and Campbell Scientific TDR100. The Acclima Digital TDT Sensor has the advantage over TDR in that signal transmitting and sampling hardware is located in the sensor head negating cable losses. TDT is also advantageous in that one-way travel time reduces signal attenuation in the sample (assuming sensor rods are the same length). Although the Acclima Digital TDT Sensor is presently geared for closed-loop irrigation control, where excavation is necessary for installation, refinement of the rod geometry for insertion (and perhaps conversion to a TDR measurement) will likely rank this TDT method alongside its TDR counterpart as an accepted laboratory and field standard for determining soil water content.

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